

Spring 2002



EEE598D: Analog Filter & Signal Processing Circuits

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Tuesday February 5, 2002



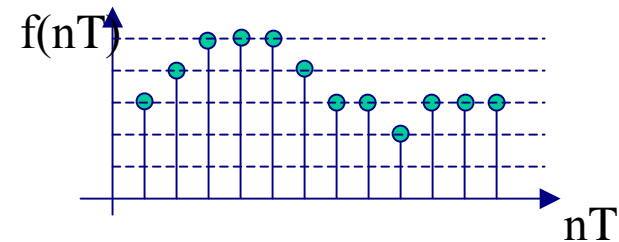
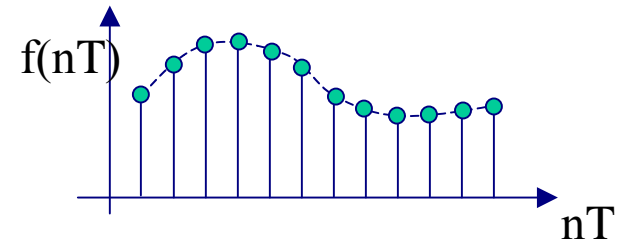
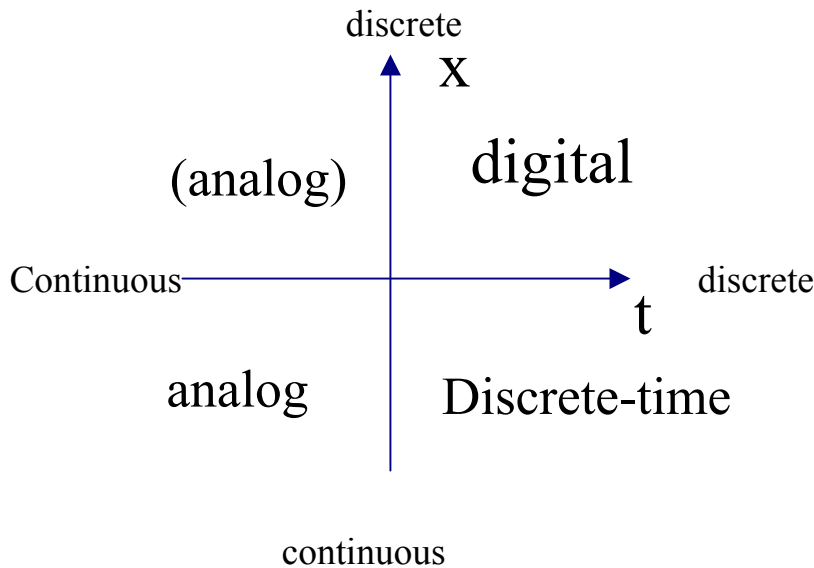
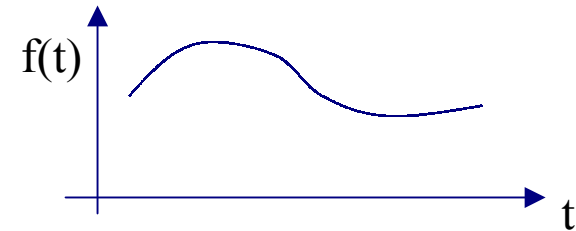
Today: Discrete-Time Filter Fundamental

- Classification of Signals
- Sampling of Signals
- Z- Transformation
- Z-Domain Transfer Function
- Block Diagram and SFG

Classification of Signals



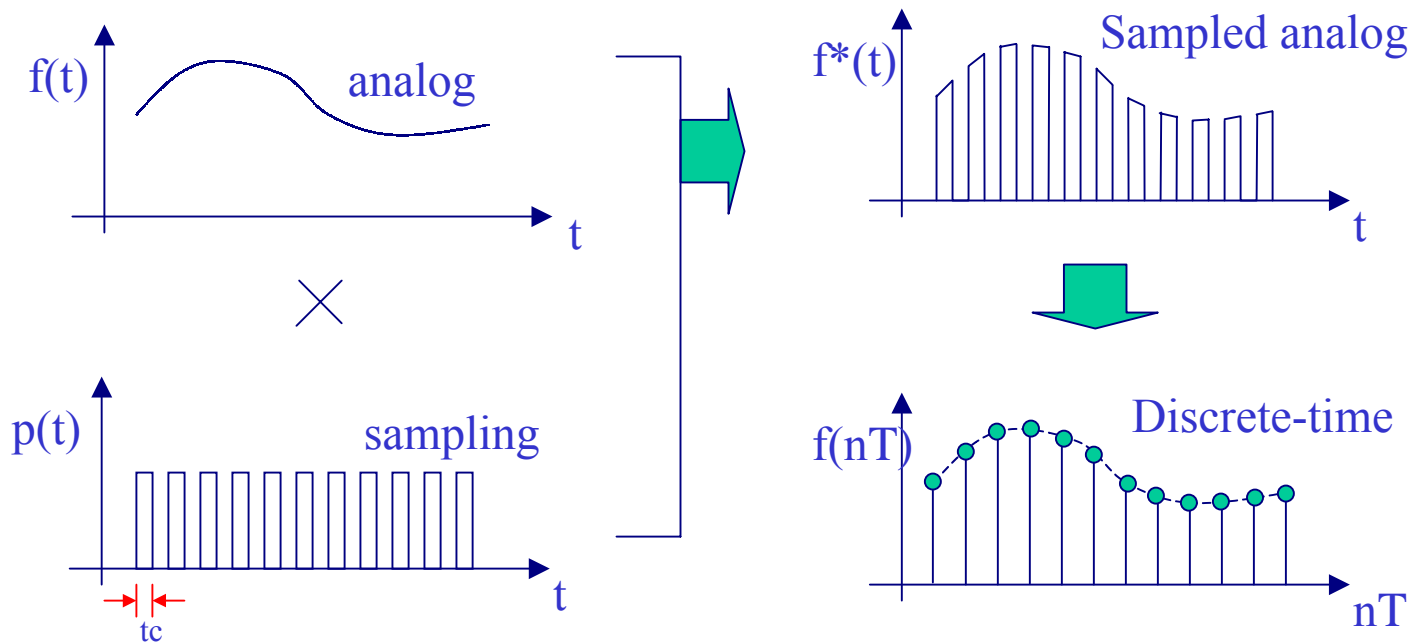
- Continuous-Time Signal
- Discrete-Time Signals
- Digital Signals



Sampling of Analog Signal



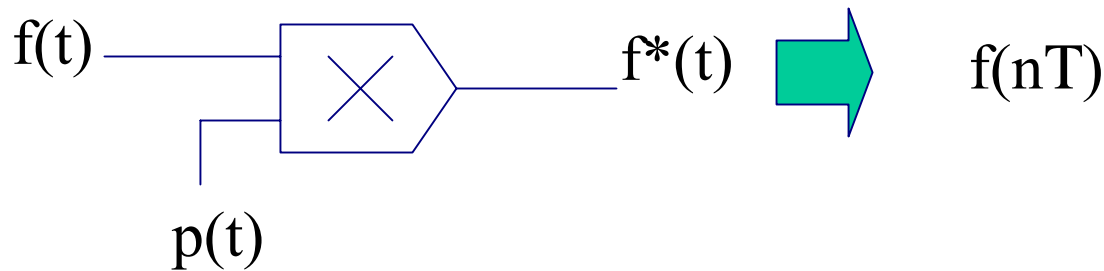
- A process converting continuous-time analog signal to discrete-time signal



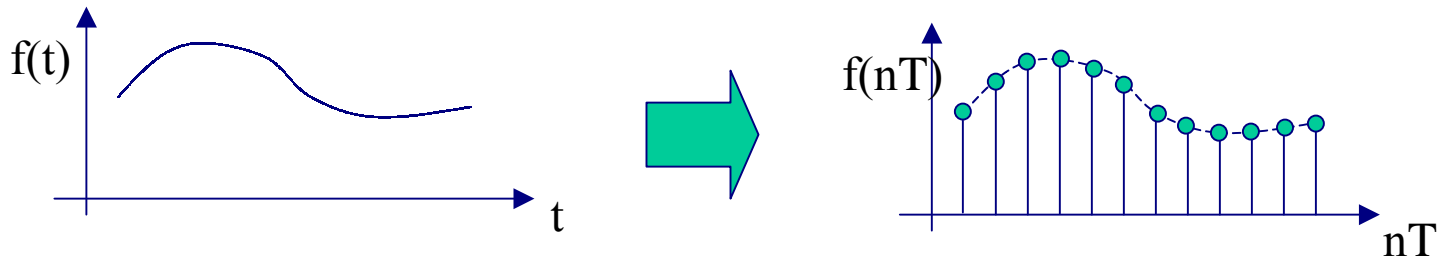
Sampling of Analog Signal



- Sampling can be modeled as the modulation of the analog signal $f(t)$ with a sampling function $p(t)$



$$f^*(t) = f(t) \cdot p(t) \quad \Rightarrow \quad f(nT)$$



Some Special Functions



- Unit Impulse Function (Dirac Delta Function):

$$\delta(t) \equiv \begin{cases} 0 & t \neq 0 \\ \infty & t = 0 \end{cases} \quad \& \quad \int_{-\infty}^{\infty} \delta(t) dt = 1$$

- Discrete Unit Sampling Function:

$$\delta(nT) \equiv \begin{cases} 0 & n \neq 0 \\ 1 & n = 0 \end{cases}$$

- Unit Step Function

$$u(nT) \equiv \begin{cases} 0 & n < 0 \\ 1 & n \geq 0 \end{cases}$$

Ideal Sampling of Analog Signal

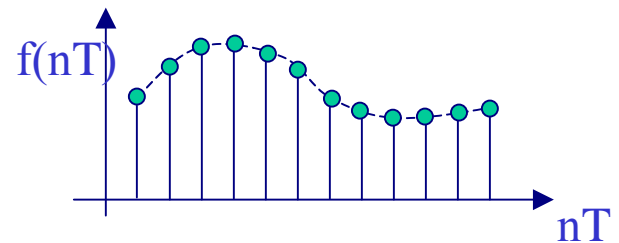
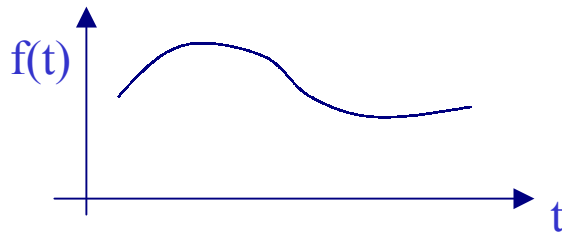


- Ideal Sampling ($t_c \rightarrow 0$):

$$f^*(t) = f(t) \cdot \sum_{n=-\infty}^{\infty} \delta(t - nT) = \sum_{n=-\infty}^{\infty} f(nT) \cdot \delta(t - nT)$$



$$f(nT) = f(t) \Big|_{t=nT} = \sum_{k=-\infty}^{\infty} f(kT) \cdot \delta((n-k)T)$$



Sampling Theory



- Sampling can be viewed as a data compression process.
- To keep useful information after compression, sampling rate must be at least twice of the useful signal frequency

$$f_{sample} \geq 2 \cdot f_{signal} \equiv f_{Niquist}$$

- Signal frequency higher than 1/2 sample rate (if not interested) must be filtered out to avoid aliasing

Z-Transformation



- (One-side) Laplace Transform

$$F(s) \equiv L\{f(t)\} \equiv \int_0^{\infty} f(t)e^{-st} dt$$

- (One-side) Z-Transform

$$F(z) \equiv Z\{f(nT)\} \equiv \sum_{n=0}^{\infty} f(nT) \cdot z^{-n}$$

Useful Z-Transform Pairs



$f(nT)$	$F(z)$
$f(nT - kT), k = 0, 1, 2, \dots$	$z^{-k}F(z)$
$f(nT + kT), k = 1, 2, \dots$	$z^kF(z) - \sum_{m=0}^{k-1} f(mT)z^{-m}$
$\sum_{m=0}^{n-1} f(mT)$	$(z - 1)^{-1}F(z)$
$a^{-n}f(nT)$	$F(az)$
$nf(nT)$	$-z \frac{dF(z)}{dz}$
$f(-nT)$	$F(1/z)$
$\sum_{k=0}^n f_1(kT)f_2(nT - kT)$	$F_1(z)F_2(z)$
$k_1f_1(nT) + k_2f_2(nT)$	$k_1F_1(z) + k_2F_2(z)$

Useful Z-Transform Pairs

$$\delta(nT)$$

$$1$$

$$Ku(nT), K$$

$$\frac{Kz}{z-1}$$

$$a^{nT}$$

$$\frac{z}{z-a^T}$$

$$a^{nT} \sin n\omega_0 T$$

$$\frac{(a^T \sin \omega_0 T) z}{z^2 - (2a^T \cos \omega_0 T) z + a^{2T}}$$

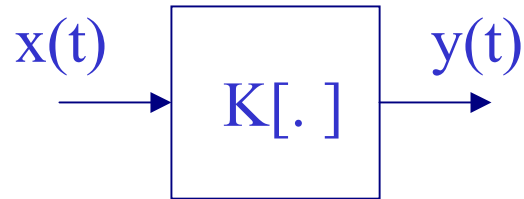
$$a^{nT} \cos n\omega_0 T$$

$$\frac{z(z - a^T \cos \omega_0 T)}{z^2 - (2a^T \cos \omega_0 T) z + a^{2T}}$$

$$nT a^{nT-T}$$

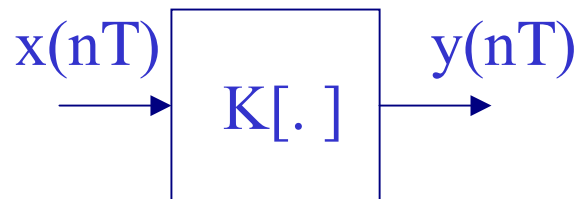
$$\frac{zT}{(z - a^T)^2}$$

System Model of Filters



$$y(t) = K(x(t))$$

A) Continuous-Time System



$$y(nT) = K(x(nT))$$

B) Discrete-Time System

Linear Time Invariant (LTI) Models



- Linear Systems

$$K(a_1x_1(t) + a_2x_2(t)) = a_1K(x_1(t)) + a_2K(x_2(t))$$

Diagram showing the linearity property: two pink arrows labeled "constant" point from the terms a_1 and a_2 in the right-hand side of the equation to the corresponding coefficients in the left-hand side.

$$K(a_1x_1(n\tau) + a_2x_2(n\tau)) = a_1K(x_1(n\tau)) + a_2K(x_2(n\tau))$$

- Time- (shift-) Invariant Systems

$$\text{if } y(t) = K(x(t)) \quad \text{Then } y(t - t_o) = K(x(t - t_o))$$

$$\text{if } y(n\tau) = K(x(nT)) \quad \text{Then } y((n - k)T) = K(x((n - k)T))$$

S-Domain Representation of (LTI) Systems



$$L\{y(t)\} = L\{K(x(t))\} = L\left\{K\left(\int_{-\infty}^{\infty} x(\zeta)\delta(t-\zeta)d\zeta\right)\right\}$$

$$= L\left\{\int_{-\infty}^{\infty} x(\zeta)K(\delta(t-\zeta))d\zeta\right\} = L\{x(t)\} \cdot L\{K(\delta(t))\}$$

$$Y(s) \equiv L\{y(t)\}$$

$$X(s) \equiv L\{x(t)\}$$

Let:

$$h(t) \equiv K(\delta(t))$$

$$H(s) \equiv L\{h(t)\}$$

Impulse response of system

Then:

$$\frac{Y(s)}{X(s)} = H(s)$$

Transfer Function (TF) of system

Note: Continuous-time LTI System can be completely determined by its impulse response.

Z-Domain Representation of (LTI) Systems



$$\begin{aligned} Z\{y(nT)\} &= Z\{K(x(nT))\} = Z\left\{K\left(\sum_{k=-\infty}^{\infty} x(kT)\delta((n-k)T)\right)\right\} \\ &= Z\left\{\sum_{k=-\infty}^{\infty} x(kT)K(\delta((n-k)T))\right\} = Z\{x(nT)\} \cdot Z\{K(\delta(nT))\} \end{aligned}$$

$$Y(z) \equiv Z\{y(nT)\}$$

Let: $X(z) \equiv Z\{x(nT)\}$

$$h(n\tau) \equiv K(\delta(nT))$$

$$H(z) \equiv Z\{h(nT)\}$$

Impulse response of system

Then:

$$\frac{Y(z)}{X(z)} = H(z)$$

Transfer Function (TF) of system

Note: Discrete-time LTI System can be completely determined by its unit sample function response.

Z-Domain Transfer Function of the System



- In general a DT system can be expressed as a linear difference equation:

$$\sum_{i=0}^m a_i y((n-i)T) = \sum_{i=0}^k b_i x((n-i)T)$$

- Z-domain transfer function of the system is defined as

$$H(z) = \frac{Y(z)}{X(z)} = \frac{b_k z^{-k} + b_{k-1} z^{-(k-1)} + \dots + b_0}{a_m z^{-m} + a_{m-1} z^{-(m-1)} + \dots + 1}$$

Steady State Response of CT LTI Systems



$$x(t) = Au(t)e^{j\omega t} \Rightarrow X(s) = A \int_0^{\infty} e^{-(s-j\omega)t} dt = \frac{A}{s-j\omega}$$

$$Y(s) = H(s) \cdot X(s) = \frac{H(s)}{s-j\omega} = \frac{H(j\omega)}{s-j\omega} + \text{Other Terms}$$

$$y(t)\Big|_{t \rightarrow \infty} = Au(t)H(j\omega)e^{j\omega t} + y_h(t) = H(j\omega)x(t)$$

Approach zero for stable system

Steady State Response of the system

Steady State Response of DT LTI Systems



$$x(t) = Au(t)e^{j\omega t} \Rightarrow X(z) = \sum_{n=0}^{\infty} Ae^{j\omega T} z^{-n} = \frac{A}{1 - z^{-1}e^{j\omega T}}$$

$$Y(z) = \frac{AH(z)}{1 - z^{-1}e^{j\omega T}} = \frac{AH(e^{j\omega T})}{1 - z^{-1}e^{j\omega T}} + Y_h(z)$$

$$y(nT) \Big|_{n \rightarrow \infty} = Au(t)H(e^{j\omega T})e^{j\omega Tn} + y_h(t) = H(e^{j\omega T})x(nT)$$

Approach zero for stable system

Steady State Response of the system

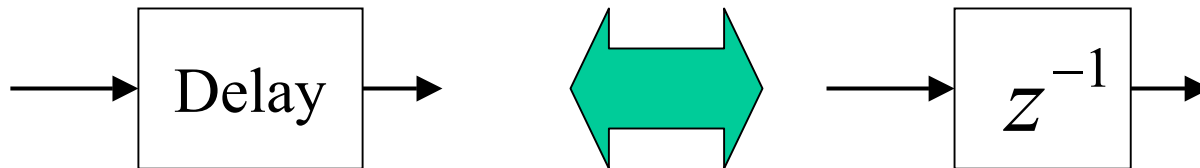
Frequency Response of the System



- System frequency response can be calculated by replace Z by EXP(j ω T):

$$H(z) \Big|_{z=e^{j\omega T}}$$

Mapping From Real Space To S-Domain



Block Diagram & Signal-Flow Graph



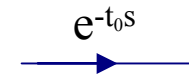
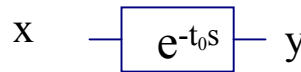
- Each Continuous-Time or Discrete-Time System Can be represented by its
 - Schematic
 - Differential or Difference Equations
 - Transfer Function
 - Block Diagram, or
 - Signal-Flow Graph

Block Diagram & Signal-Flow Graph

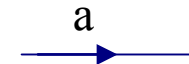
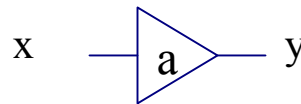


- Basic Elements (Continuous-Time)

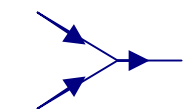
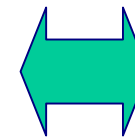
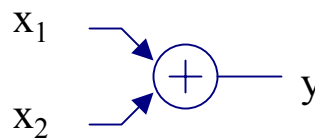
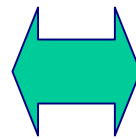
$$y(t) = x(t - t_0)$$



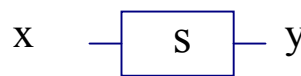
$$y(t) = ax(t)$$



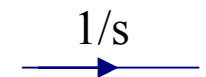
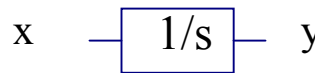
$$y(t) = x_1(t) + x_2(t)$$



$$y(t) = \frac{dx(t)}{dt}$$



$$y(t) = y(0) + \int_0^t x(t_1) dt_1$$



Block Diagram & Signal-Flow Graph

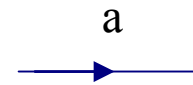
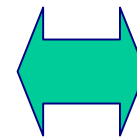
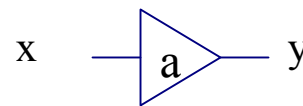
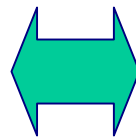


- Basic Elements (Discrete-Time)

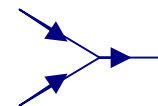
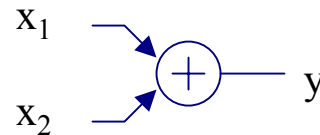
$$y(nT) = x((n-1)T)$$



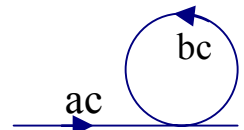
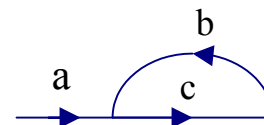
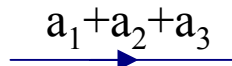
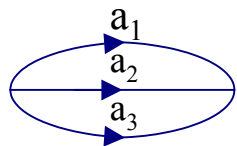
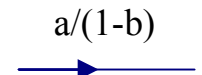
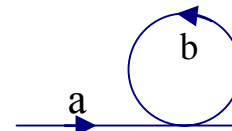
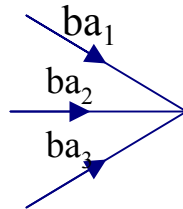
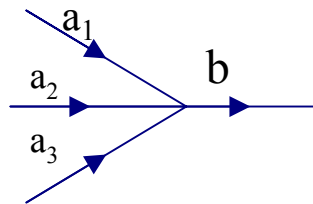
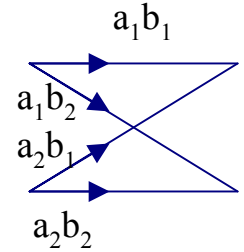
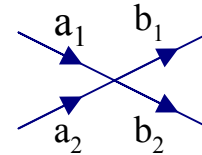
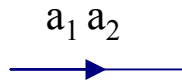
$$y(nT) = ax(nT)$$



$$y(nT) = x_1(nT) + x_2(nT)$$



Rules for Signal-Flow Graph Reduction

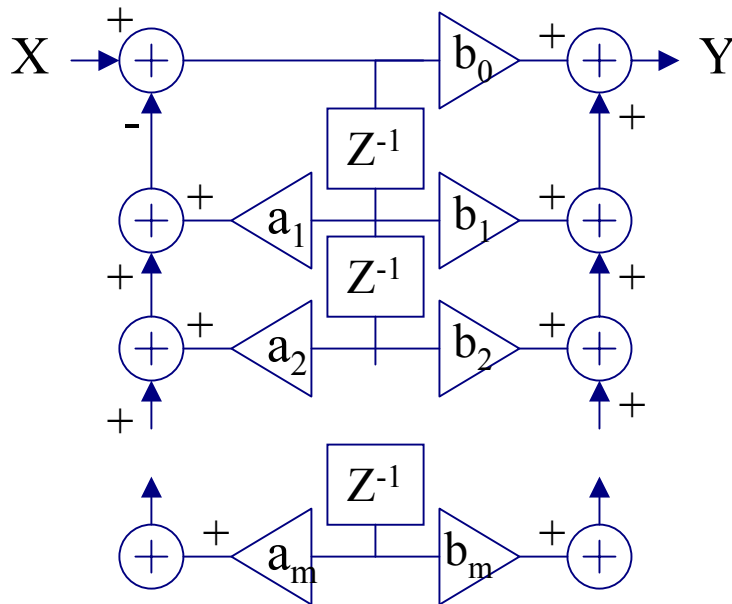


Block Diagram & Signal-Flow Graph



- Example

$$y(nT) + \sum_{i=1}^m a_i y((n-i)T) = \sum_{i=0}^m b_i x((n-i)T)$$



or

