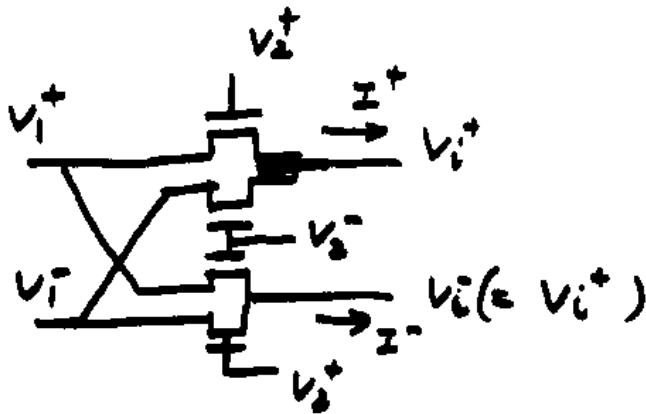


HW#2

problem 1 solution



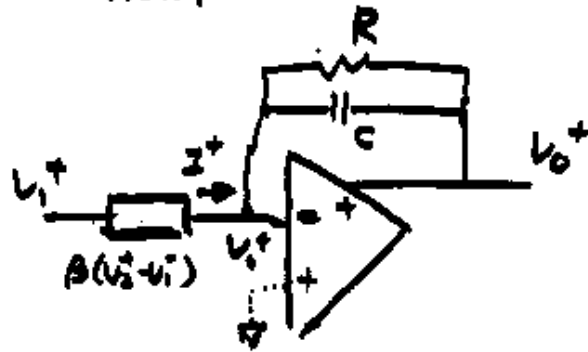
For the 4-T VCR, we have

$$\begin{aligned} I^+ - I^- &= \beta \left[ v_2^+ - v_T - \frac{v_1^+ + v_i^+}{2} \right] [v_1^+ - v_i^+] \\ &\quad + \beta \left[ v_2^- - v_T - \frac{v_1^- + v_i^-}{2} \right] [v_1^- - v_i^-] \\ &\quad - \beta \left[ v_2^+ - v_T - \frac{v_1^+ + v_i^-}{2} \right] [v_i^- - v_i^-] \\ &\quad - \beta \left[ v_2^- - v_T - \frac{v_1^+ + v_i^+}{2} \right] [v_1^+ - v_i^-] \\ &= \beta [v_2^+ - v_i^-] [v_1^+ - v_i^-] \quad \textcircled{1} \end{aligned}$$

Since  $v_{cm}$  (common mode voltage) = 0

$$\begin{cases} I^+ = -I^- \\ v_1^+ = -v_1^- \\ v_o^+ = v_o^- \\ v_i^+ = v_i^- \end{cases} \Rightarrow \begin{cases} I^+ - I^- = 2I^+ \\ v_1^+ - v_1^- = 2v_1^+ \\ v_o^+ - v_o^- = 2v_o^+ \\ v_i^+ = v_i^- = 0 \end{cases} \quad \textcircled{2}$$

we may use half circuit to simplify the calculation:

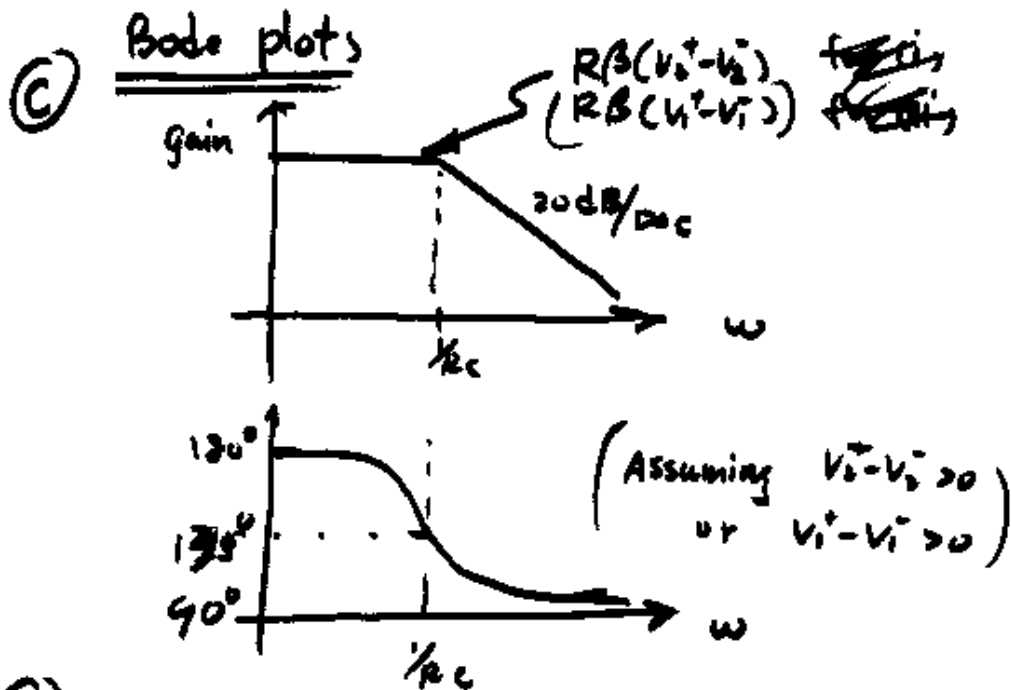


$$\begin{aligned} \underline{v_o^+ - v_o^-} &= -\beta [v_o^+ - v_o^-] [v_i^+ - v_i^-] [R \parallel \frac{1}{sC}] \\ &= -\beta (v_o^+ - v_o^-) (v_i^+ - v_i^-) \frac{1}{\frac{1}{R} + s} \end{aligned}$$

$$\textcircled{a} \quad \frac{v_o^+ - v_o^-}{v_i^+ - v_i^-} \Big|_{v_o^+ - v_o^- = \text{const.}} = \frac{-\beta (v_o^+ - v_i^-) \cdot \frac{1}{C}}{\frac{1}{RC} + s}$$

$$\frac{v_o^+ - v_o^-}{v_o^+ - v_o^-} \Big|_{v_i^+ - v_i^- = \text{const.}} = -\frac{\beta (v_o^+ - v_i^-) \frac{1}{C}}{\frac{1}{RC} + s}$$

⑥ no finite zero. in  $TF$   
 single pole at  $\underline{s = -\frac{1}{RC}}$

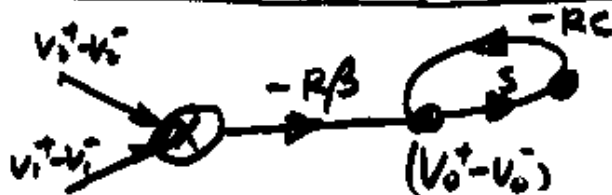


(d) SFG

$$(V_o^+ - V_o^-) = -\frac{\beta(V_o^- - V_o^+)(V_i^+ - V_i^-)}{RC + s}$$

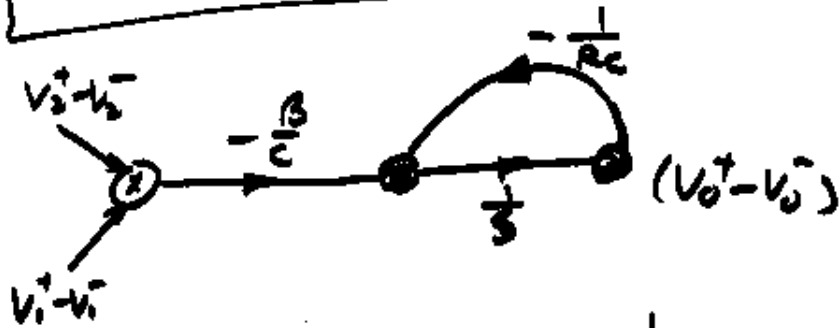
$$\Rightarrow (RC + s)(V_o^+ - V_o^-) = -\beta(V_o^- - V_o^+)(V_i^+ - V_i^-)$$

$$(V_o^+ - V_o^-) = -R\beta(V_o^- - V_o^+)(V_i^+ - V_i^-) - RCs(V_o^+ - V_o^-)$$

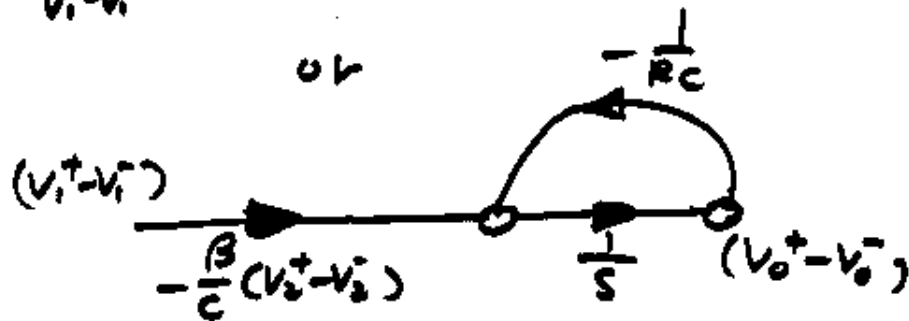


or in integrator version

$$(V_o^+ - V_o^-) = -\frac{\beta}{c} (V_2^+ - V_2^-) (V_1^+ - V_1^-) \frac{1}{s} - \frac{1}{Rc} (V_o^+ - V_o^-) \frac{1}{s}$$

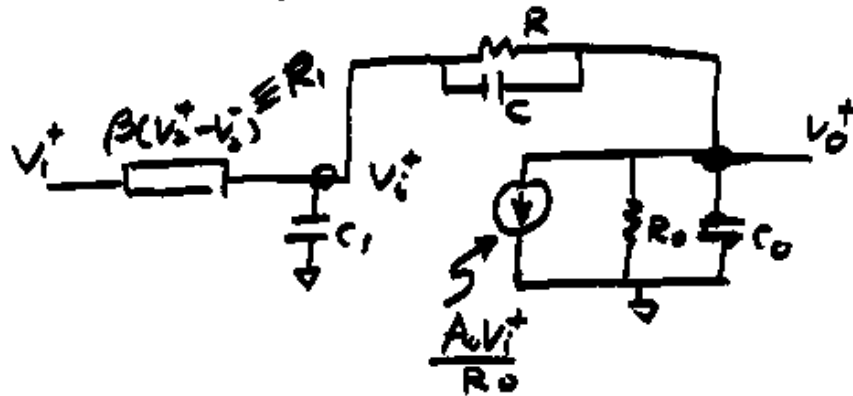


or



## problem 2 Solution

Assuming  $V_i^+$  &  $V_i^-$  still very close, the 4-T VCR equation (1) still valid. We have the half equivalent circuit w/ finite op gain & bandwidth,



We have:

$$\Sigma I = 0 \text{ @ } V_i^+:$$

$$(A) \quad V_i^+ \left[ \frac{1}{R_1} + sC_1 + \frac{1}{R} + sC \right] = V_i^+ \frac{1}{R_1} + V_o^+ \left[ \frac{1}{R} + sC \right]$$

$$\Sigma I = 0 \text{ @ } V_o^+:$$

$$(B) \quad V_o^+ \left[ \frac{1}{R} + sC + \frac{1}{R_0} + sC_0 \right] = V_i^+ \left[ \frac{1}{R} + sC \right] - \frac{A_0}{R_0} V_i^+$$

From (B)

$$V_i^+ = -V_o^+ \frac{\frac{1}{R} + \frac{1}{R_o} + (C + C_o)S}{\frac{A_o}{R_o} - (\frac{1}{R} + CS)} \quad (C)$$

insert (C) into (A)

$$-V_o^+ \frac{(\frac{1}{R} + \frac{1}{R_o} + (C + C_o)S)(\frac{1}{R_1} + \frac{1}{R} + S(C_1 + C))}{\frac{A_o}{R_o} - (\frac{1}{R} + CS)}$$

$$= V_i^+ \frac{1}{R_1} + V_o^+ [\frac{1}{R} + CS]$$

$$\Rightarrow -V_o^+ \left[ \frac{(\frac{1}{R} + \frac{1}{R_o} + (C + C_o)S)(\frac{1}{R_1} + \frac{1}{R} + S(C_1 + C))}{\frac{A_o}{R_o} - (\frac{1}{R} + CS)} + \frac{1}{R} + CS \right]$$

$$= V_i^+ \frac{1}{R_1}$$

$$\Rightarrow \frac{V_o^+}{V_i^+} = - \frac{\frac{1}{R_1}}{\frac{(\frac{1}{R} + \frac{1}{R_o} + (C + C_o)S)(\frac{1}{R_1} + \frac{1}{R} + S(C_1 + C))}{\frac{A_o}{R_o} - (\frac{1}{R} + CS)} + \frac{1}{R} + CS}$$

$$\Rightarrow \frac{V_0^+}{V_i^+} = - \frac{\left(\frac{A_0}{R_0} - \left(\frac{1}{R} + Cs\right)\right) \frac{1}{R_1}}{\left[\left(\frac{1}{R} + \frac{1}{R_0}\right) + (C + C_0)s\right] \left[\frac{1}{R_1} + \frac{1}{R} + s(C + C_0)\right]}$$

$$\hookrightarrow \frac{A_0}{R_0} \left(\frac{1}{R} + Cs\right) - \left(\frac{1}{R} + Cs\right)^2$$

$$\boxed{\frac{V_0^+}{V_i^+} = - \frac{K (z - s)}{s^2 + as + b}} \quad \text{is a 2<sup>nd</sup> order filter}$$

where  $K$ ,  $z$ ,  $a$ ,  $b$ , are constants;

There some special cases:

① at low frequency,  $\beta$  high  $A_0$

$$\frac{V_0^+}{V_i^+} = \frac{V_0^+ - V_0^-}{V_i^+ - V_i^-} = - \frac{\frac{A_0}{R_0} \frac{1}{R_1}}{\frac{A_0}{R_0} \left(\frac{1}{R} + Cs\right)} = - \frac{\beta (V_0^+ - V_0^-) \frac{1}{R}}{\frac{1}{R} + Cs}$$

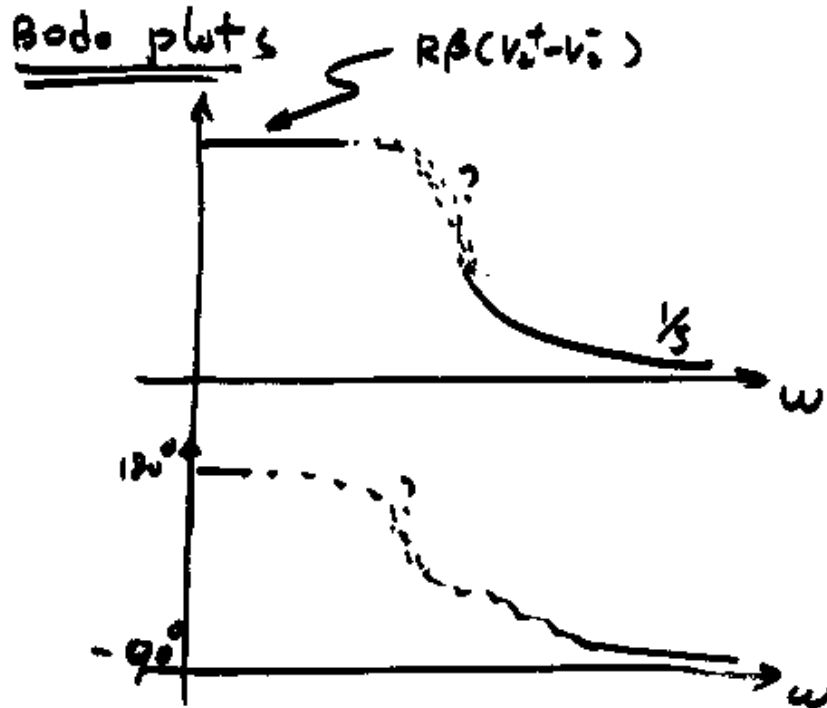
same as ideal op.

this is used to low frequency band of the mixer.

② at high frequency

$$\frac{V_o^+}{V_i^+} \approx \frac{V_o^+ - V_o^-}{V_i^+ - V_i^-} \approx \frac{\frac{C}{R_1} s}{(C + C_0)(C_1 + C) \approx C^2} \frac{1}{s} \rightarrow 0$$

the mixer use this to filter out the high frequency (2x) band.





SFG

From (A)

$$V_i^+ = V_i^+ \frac{1}{R_1(c+c_1)} \frac{1}{s} + V_o^+ \left[ \frac{1}{R(c+c_1)} \frac{1}{s} + \frac{c}{c+c_1} \right] - V_i^+ \left( \frac{1}{R_1} + \frac{1}{R} \right) \frac{1}{c+c_1} \cdot \frac{1}{s}$$

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From (B)

$$V_o^+ = \frac{V_i^+ \left[ \frac{1}{R} + cs - \frac{A_o}{R_o} \right] - \left( \frac{1}{R} + \frac{1}{R_o} \right) V_o^+}{(c+c_o)s}$$
$$= V_i^+ \left[ \left( \frac{1}{R} - \frac{A_o}{R_o} \right) \frac{1}{c+c_o} \cdot \frac{1}{s} + \frac{c}{c+c_o} \right] - \frac{\frac{1}{R} + \frac{1}{R_o}}{c+c_o} \frac{1}{s} V_o^+$$

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